

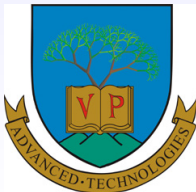
Számítógépvezérelt szabályozások elmélete

Stabilitás

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2011. március 14.



Overview

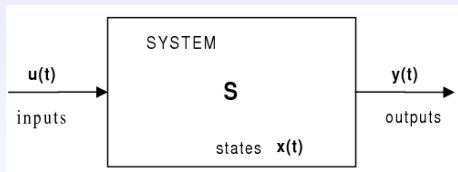
- 1 Basic notions
- 2 Bounded input-bounded output (BIBO) stability
- 3 Stability in the state space
- 4 Stability region of nonlinear systems

Systems

- System (**S**): acts on signals

$$y = \mathbf{S}[u]$$

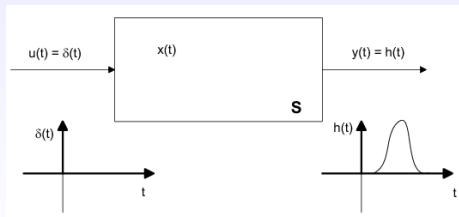
- inputs (u) and outputs (y)



CT-LTI I/O system models

- Time domain: **Impulse response function** is the response of a SISO LTI system to a Dirac-delta input function with zero initial condition.
- The output of **S** can be written as

$$y(t) = \int_{-\infty}^{\infty} h(t - \tau)u(\tau)d\tau = \int_{-\infty}^{\infty} h(\tau)u(t - \tau)d\tau$$



CT-LTI state-space models

- General form - revisited

$$\begin{aligned}\dot{x}(t) &= Ax(t) + Bu(t) \quad , \quad x(t_0) = x(0) \\ y(t) &= Cx(t)\end{aligned}$$

with

- signals: $x(t) \in \mathbb{R}^n$, $y(t) \in \mathbb{R}^p$, $u(t) \in \mathbb{R}^r$
- system parameters: $A \in \mathbb{R}^{n \times n}$, $B \in \mathbb{R}^{n \times r}$, $C \in \mathbb{R}^{p \times n}$ ($D = 0$)
- Dynamic system properties:
 - observability ✓
 - controllability ✓
 - stability

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Signal spaces

- L_q spaces

$$L_q[0, \infty[= \left\{ f : \mathbb{R}_0^+ \mapsto \mathbb{R} \mid f \text{ is measurable and } \int_0^\infty |f(t)|^q < \infty \right\}$$

special case

$$L_\infty[0, \infty[= \left\{ f : \mathbb{R}_0^+ \mapsto \mathbb{R} \mid f \text{ is measurable and } \sup_{t \in \mathbb{R}_0^+} |f(t)| < \infty \right\}$$

- Remark: L_q spaces are Banach spaces with norms

$$\|f\|_q = \left(\int_0^\infty |f(t)|^q \right)^{1/q}$$

$$\|f\|_\infty = \sup_{t \in \mathbb{R}_0^+} |f(t)|$$

Vector valued signals

- $f(t) \in \nu, \forall t \geq 0$, where ν is a finite dimensional linear space (e.g. $\nu = \mathbb{R}^n$)
- $\|\cdot\|_\nu$ is a norm in ν (e.g. Euclidean)

$$L_q(\nu) = \left\{ f : \mathbb{R}_0^+ \mapsto \nu \mid f \text{ is measurable and } \int_0^\infty \|f(t)\|_\nu^q < \infty \right\}$$

$$\text{norm: } \|f\|_q = \left(\int_0^\infty \|f(t)\|_\nu^q \right)^{\frac{1}{q}}$$

- Remark: The case L_2 is special, because the norm comes from an inner product (L_2 is a Hilbert-space)

BIBO stability – general

Definition

A system is **externally or BIBO stable** if for any bounded input it responds with a bounded output

$$\|u\| \leq M_1 < \infty \Rightarrow \|y\| \leq M_2 < \infty$$

where $\|\cdot\|$ is a signal norm.

- This applies to any type of systems.
- **Stability is a system property**, i.e. it is realization-independent.

BIBO stability – 1

- Bounded input-bounded output (BIBO) stability for SISO systems

$$|u(t)| \leq M_1 < \infty, \forall t \in [0, \infty[\Rightarrow |y(t)| \leq M_2 < \infty, \forall t \in [0, \infty[$$

Theorem (BIBO stability)

A SISO LTI system is BIBO stable if and only if

$$\int_0^{\infty} |h(t)| dt \leq M < \infty$$

where $M \in \mathbb{R}^+$ and $h(t)$ is the impulse response function.

- Question: How can we generalize for MIMO systems?

BIBO stability – 2

Proof.

- \Leftarrow Assume $\int_0^\infty |h(t)| dt \leq M < \infty$ and u is bounded, i.e. $|u(t)| \leq M_1 < \infty, \forall t \in \mathbb{R}_0^+$. Then

$$|y(t)| \leq \left| \int_0^\infty h(\tau) u(t - \tau) d\tau \right| \leq M_1 \int_0^\infty |h(\tau)| d\tau \leq M_1 \cdot M = M_2$$

- \Rightarrow (indirect) Assume $\int_0^\infty |h(\tau)| d\tau = \infty$, but the system is BIBO stable. Consider the **bounded** input:

$$u(t - \tau) = \text{sign } h(\tau) = \begin{cases} 1 & \text{if } h(\tau) > 0 \\ 0 & \text{if } h(\tau) = 0 \\ -1 & \text{if } h(\tau) < 0 \end{cases}$$

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Stability of nonlinear systems

- Consider the **autonomous** nonlinear system:

$$\dot{x} = f(x), \quad x \in \mathcal{X} = \mathbb{R}^n, \quad f : \mathbb{R}^n \mapsto \mathbb{R}^n$$

with an equilibrium point: $f(x^*) = 0$

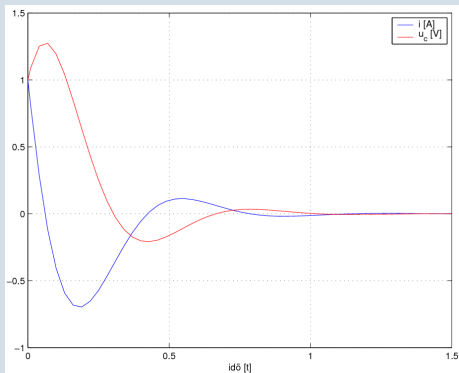
- x^* **stable equilibrium point**: for any $\epsilon > 0$ there exists $\delta > 0$ ($\delta < \epsilon$) such that for $\|x^* - x(0)\| < \delta$ $\|x^* - x(t)\| < \epsilon$ holds.
- x^* **asymptotically stable equilibrium point**: x^* stable and $\lim_{t \rightarrow \infty} x(t) = x^*$.
- x^* **unstable equilibrium point**: not stable
- x^* **locally (asymptotically) stable**: there exists a neighborhood U of x^* within which the (asymptotic) stability conditions hold
- x^* **globally (asymptotically) stable**: $U = \mathbb{R}^n$

Asymptotic stability

Example (Asymptotically stable system)

RLC circuit with parameters: $R = 1 \Omega$, $L = 10^{-1}H$, $C = 10^{-1}F$.

$u_C(0) = 1 \text{ V}$, $i(0) = 1 \text{ A}$, $u_{be}(t) = 0 \text{ V}$

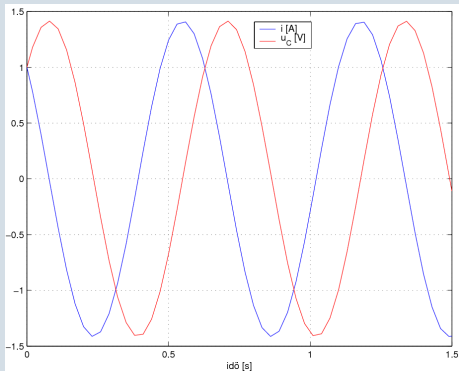


Non-asymptotic stability

Example (Non-asymptotically stable system)

RLC circuit with parameters: $R = 0 \Omega$, $L = 10^{-1} H$, $C = 10^{-1} F$.

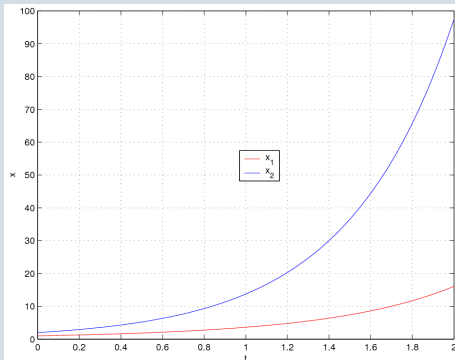
$u_C(0) = 1 \text{ V}$, $i(0) = 1 \text{ A}$, $u_{be}(t) = 0 \text{ V}$



Instability

Example (Unstable system)

$$\begin{aligned}\dot{x}_1 &= x_1 + 0.1x_2 \\ \dot{x}_2 &= -0.2x_1 + 2x_2\end{aligned}, \quad x(0) = \begin{bmatrix} 1 \\ 2 \end{bmatrix}$$



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Stability of CT-LTI systems

- Truncated state equation with $u \equiv 0$:

$$\dot{x} = A \cdot x, \quad x \in \mathbb{R}^n, \quad A \in \mathbb{R}^{n \times n}, \quad x(0) = x_0$$

- Equilibrium point: $x^* = 0$
- Solution:

$$x(t) = e^{At} \cdot x_0$$

- Recall:** A can be diagonalized (there exists invertible T , such that

$$T \cdot A \cdot T^{-1}$$

is diagonal) if and only if, A has n linearly independent eigenvectors.

Asymptotic stability of LTI systems – 1

Stability types:

- the real part of every eigenvalue of A is negative (A is a *stability matrix*): **asymptotic stability**
- A has eigenvalues with zero and negative real parts
 - the eigenvectors related to the zero real part eigenvalues are linearly independent: **(non-asymptotic) stability**
 - the eigenvectors related to the zero real part eigenvalues are not linearly independent: **(polynomial) instability**
- A has (at least) an eigenvalue with positive real part: **(exponential) instability**

Asymptotic stability of LTI systems – 2

Theorem

The eigenvalues of a square $A \in \mathbb{R}^{n \times n}$ matrix remain unchanged after a similarity transformation on A by a transformation matrix T :

$$A' = TAT^{-1}$$

Proof.

Let us start with the eigenvalue equation for matrix A

$$A\xi = \lambda\xi, \quad \xi \in \mathbb{R}^n, \quad \lambda \in \mathbb{C}$$

If we transform it using $\xi' = T\xi$ then we obtain

$$TAT^{-1}T\xi = \lambda T\xi$$

$$A'\xi' = \lambda\xi'$$

Asymptotic stability of LTI systems – 3

Theorem

A CT-LTI system is asymptotically stable iff A is a stability matrix.

Sketch of the Proof.

Assume, that A is diagonalizable

$$\bar{A} = TAT^{-1} = \begin{bmatrix} \lambda_1 & 0 & \dots & 0 \\ 0 & \lambda_2 & \dots & 0 \\ & & \ddots & 0 \\ 0 & \dots & 0 & \lambda_n \end{bmatrix}$$

$$\bar{x}(t) = e^{\bar{A}t} \cdot \bar{x}_0, \quad e^{\bar{A}t} = \begin{bmatrix} e^{\lambda_1 t} & 0 & \dots & 0 \\ 0 & e^{\lambda_2 t} & \dots & 0 \\ & & \ddots & 0 \\ 0 & \dots & 0 & e^{\lambda_n t} \end{bmatrix}$$



BIBO and asymptotic stability

Theorem

Asymptotic stability implies BIBO stability for LTI systems.

Proof.

$$x(t) = e^{At}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau, \quad y(t) = Cx(t)$$

$$\begin{aligned}\|x(t)\| &\leq \|e^{At}x(t_0) + M \int_0^t e^{A(t-\tau)}Bd\tau\| \\ &= \|e^{At}(x(t_0) + M \int_0^t e^{-A\tau}Bd\tau)\| \\ &= \|e^{At}(x(t_0) + M[-A^{-1}e^{-A\tau}B]_0^t)\| \\ &= \|e^{At}(x(t_0) - MA^{-1}e^{-At}B + MA^{-1}B)\| \\ \|x(t)\| &\leq \|e^{At}(x(t_0) + MA^{-1}B) - MA^{-1}B\|\end{aligned}$$



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Lyapunov theorem of stability

- **Lyapunov-function:** $V : \mathcal{X} \mapsto \mathbb{R}$
 - $V > 0$, if $x \neq x^*$, $V(x^*) = 0$
 - V is continuously differentiable
 - V is non-increasing, i.e.

$$\frac{d}{dt}V(x) = \frac{\partial V}{\partial x}\dot{x} = \frac{\partial V}{\partial x}f(x) \leq 0$$

Theorem (Lyapunov stability theorem)

- *If there exists a Lyapunov function to the system $\dot{x} = f(x)$, $f(x^*) = 0$, then x^* is a stable equilibrium point.*
- *If $\frac{d}{dt}V \not\equiv 0$ then x^* is an asymptotically stable equilibrium point.*
- *If the properties of a Lyapunov function hold only in a neighborhood U of x^* , then x^* is a locally (asymptotically) stable equilibrium point.*

Lyapunov theorem – example

Example

- System:

$$\dot{x} = -(x - 1)^3$$

- Equilibrium point: $x^* = 1$
- Lyapunov function: $V(x) = (x - 1)^2$

$$\begin{aligned}\frac{d}{dt}V &= \frac{\partial V}{\partial x}\dot{x} = 2(x - 1) \cdot (-(x - 1)^3) = \\ &= -2(x - 1)^4 < 0\end{aligned}$$

- The system is **globally asymptotically stable**

CT-LTI Lyapunov theorem – 1

Basic notions:

- $Q \in \mathbb{R}^{n \times n}$ **symmetric matrix**: $Q = Q^T$, i.e. $[Q]_{ij} = [Q]_{ji}$ (every eigenvalue of Q is real)
- symmetric matrix Q is **positive definite** ($Q > 0$): $x^T Q x > 0, \forall x \in \mathbb{R}^n, x \neq 0$ (\Leftrightarrow every eigenvalue of Q is positive)
- symmetric matrix Q is **negative definite** $Q < 0$: $x^T Q x < 0, \forall x \in \mathbb{R}^n, x \neq 0$ (\Leftrightarrow every eigenvalue of Q is negative)

Theorem (Lyapunov criterion for LTI systems)

The state matrix (A) of an LTI system is a stability matrix if and only if there exists a positive definite symmetric matrix P for every given positive definite symmetric matrix Q such that

$$A^T P + P A = -Q$$

CT-LTI Lyapunov theorem – 2

Proof.

\Leftarrow Assume $\forall Q > 0 \exists P > 0$ such that $A^T P + PA = -Q$.

Let $V(x) = x^T P x$.

$$\frac{d}{dt} V = \dot{x}^T P x + x^T P \dot{x} = x^T (A^T P + PA) x < 0$$

\Rightarrow Assume A is a stability matrix. Then

$$P = \int_0^{\infty} e^{A^T t} Q e^{A t} dt$$

$$A^T P + PA = \int_0^{\infty} A^T e^{A^T t} Q e^{A t} dt + \int_0^{\infty} e^{A^T t} Q e^{A t} A dt = \left[e^{A^T t} Q e^{A t} \right]_0^{\infty} = -Q$$

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Quadratic stability region

- Use **quadratic Lyapunov function candidate** with a given positive definite diagonal weighting matrix Q (tuning parameter!)

$$V[x(t)] = (x - x^*)^T \cdot Q \cdot (x - x^*)$$

- Dissipativity condition gives a **conservative estimate of the stability region**

$$\frac{dV}{dt} = \frac{\partial V}{\partial x} \frac{dx}{dt} = \frac{\partial V}{\partial x} \bar{f}(x)$$

- $\bar{f}(x) = f(x)$ in the open loop case with $u = 0$
- $\bar{f}(x) = f(x) + g(x) \cdot C(x)$ in the closed-loop case where $C(x)$ is the static state feedback

Quadratic stability region

Example

- Nonlinear system

$$\begin{aligned}\dot{x}_1 &= 0.4x_1x_2 - 1.5x_1 \\ \dot{x}_2 &= -0.8x_1x_2 - 1.5x_2 + 1.5u \\ y &= x_2\end{aligned}$$

- Equilibrium point with $u^* = 7.75$

$$x^* = \begin{bmatrix} x_1^* \\ x_2^* \end{bmatrix} = \begin{bmatrix} 2 \\ 3.75 \end{bmatrix}$$

- Locally linearized system

$$\begin{aligned}\dot{\tilde{x}} &= \begin{bmatrix} 0 & 0.8 \\ -3 & -3.1 \end{bmatrix} \tilde{x} + \begin{bmatrix} 0 \\ 1.5 \end{bmatrix} \tilde{u} \\ \tilde{y} &= \begin{bmatrix} 0 & 1 \end{bmatrix} \tilde{x}\end{aligned}$$

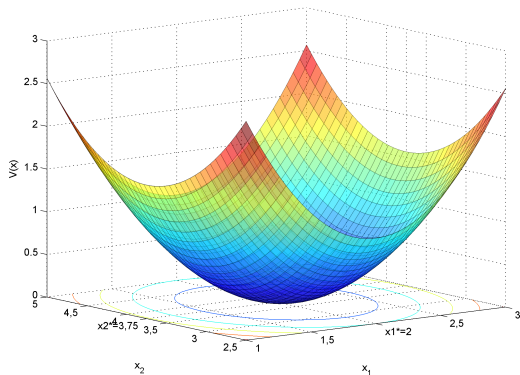
- Eigenvalues of the state matrix are $\lambda_1 = -1.5$ and $\lambda_2 = -1.6$ so equilibrium point x^* (not the system!) is locally asymptotically stable.

Quadratic stability region

Example (continued)

- Quadratic Lyapunov function

$$V(x) = (x - x^*)^T \cdot \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \cdot (x - x^*)$$



Quadratic stability region

Example (continued)

- Time derivative of the quadratic Lyapunov function

